

Chapter 2

PN junction and diodes



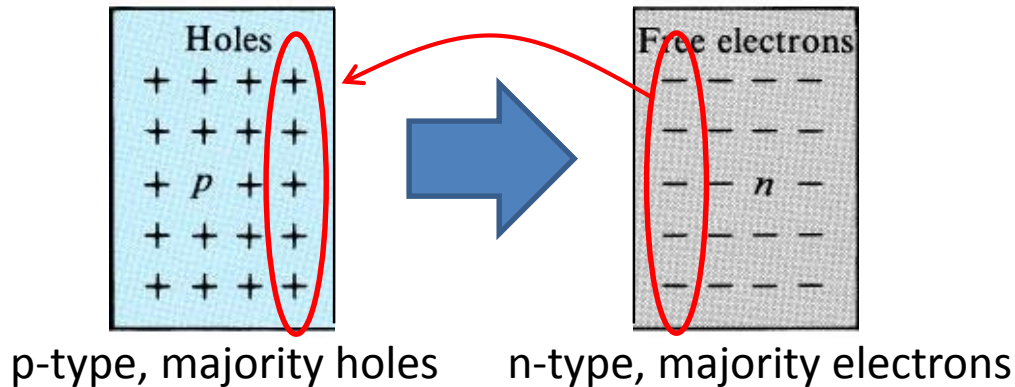
PN junction and diodes

Outline

- PN junction
 - What happens in a PN junction
 - Currents through the PN junction
 - Properties of the depletion region
- Diodes
 - Diode I/V characteristics
 - PN junction under reverse bias
 - PN junction in breakdown region
 - PN junction under forward bias
- Circuits with diodes
 - DC restorer
 - Voltage doubler
 - Voltage multiplier

PN junction

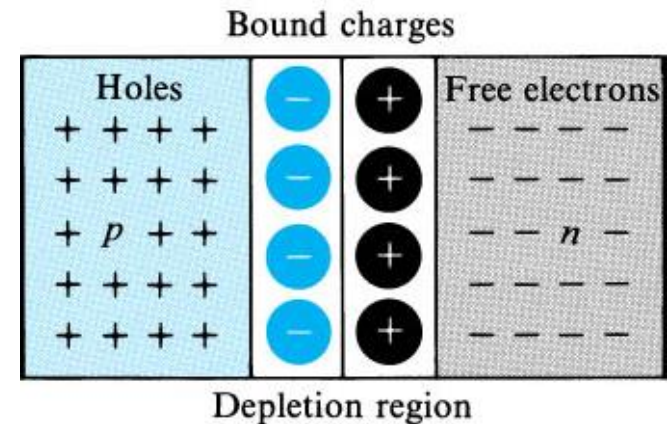
What happens at the transition region ?



At the transition region, the majority holes are neutralized by an equal amount of majority electrons

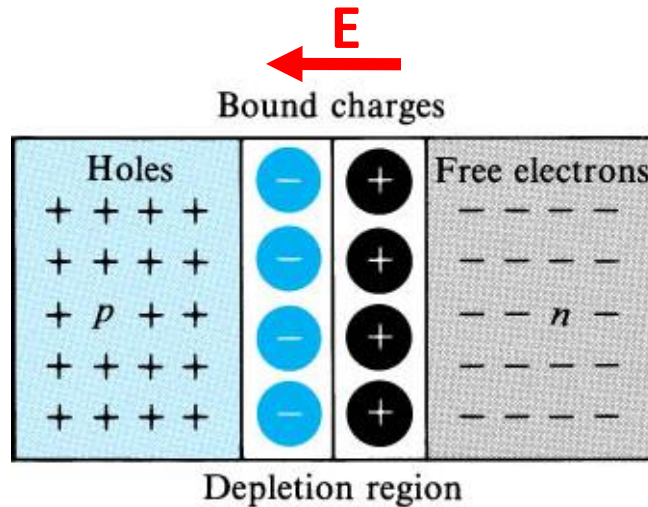
=> this reveals the bound charges (ions) that are in the crystal

=> the bound charges create an E-field



PN junction

Effect of the E-field

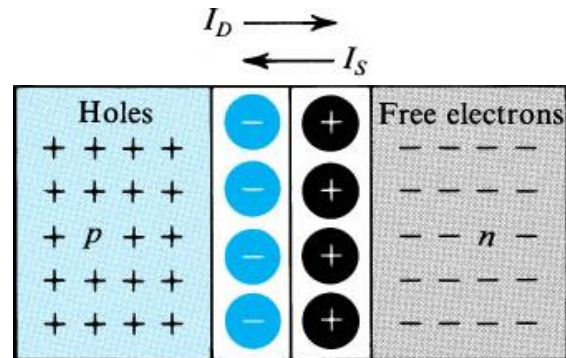


- The E-field « pushes » the holes in the p-region
- The E-field « pushes » the electrons in the n-region
- Holes/electrons need more energy to pass the depletion region
- An equilibrium appears, where the size of the depletion region no longer changes

PN junction

Currents through the depletion region

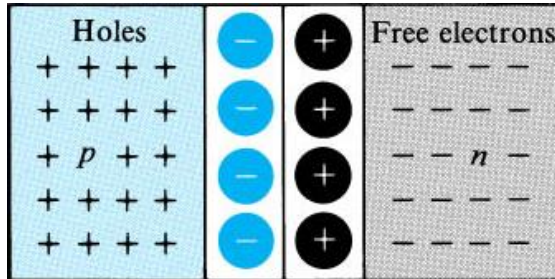
- Diffusion current I_D :
 - majority holes/electrons in the p/n-region that have enough energy to overcome the E-field
 - Diffusion current I_S :
 - minority electrons/holes in the p/n-region (created through thermal ionization)
- ⇒ Swept through the depletion region by the E-field



- No external current exists $\Rightarrow I_D = I_S$

PN junction

Width and barrier voltage of the depletion region



$$\text{div } E = \frac{\rho}{\epsilon}$$

$$E = -\text{grad } V$$

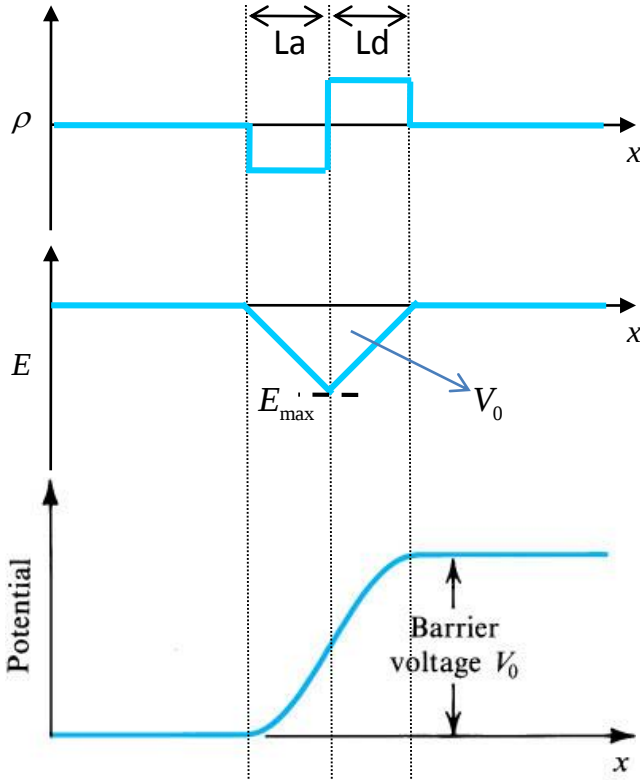
Global depletion region must be

electrically neutral $\Rightarrow N_a L_a = N_d L_d$

$$\text{div } E = \frac{\rho}{\epsilon} \Rightarrow E_{\text{max}} = \frac{q N_a L_a}{\epsilon} = \frac{q N_d L_d}{\epsilon}$$

$$E = -\text{grad } V \Rightarrow V_0 = E_{\text{max}} \frac{L_a + L_d}{2} = V_T \ln \left(\frac{N_a N_d}{n_i^2} \right)$$

$$L_a + L_d = \sqrt{\frac{2\epsilon}{q} V_0 \left(\frac{1}{N_a} + \frac{1}{N_d} \right)}$$



PN junction

Width and barrier voltage of the depletion region

- For silicon at room temperature:

$$V_0 \approx 0.6 - 0.8V$$

- Typically, the depth of the depletion region ranges from 0.1 μm to 1 μm

PN junction and diodes

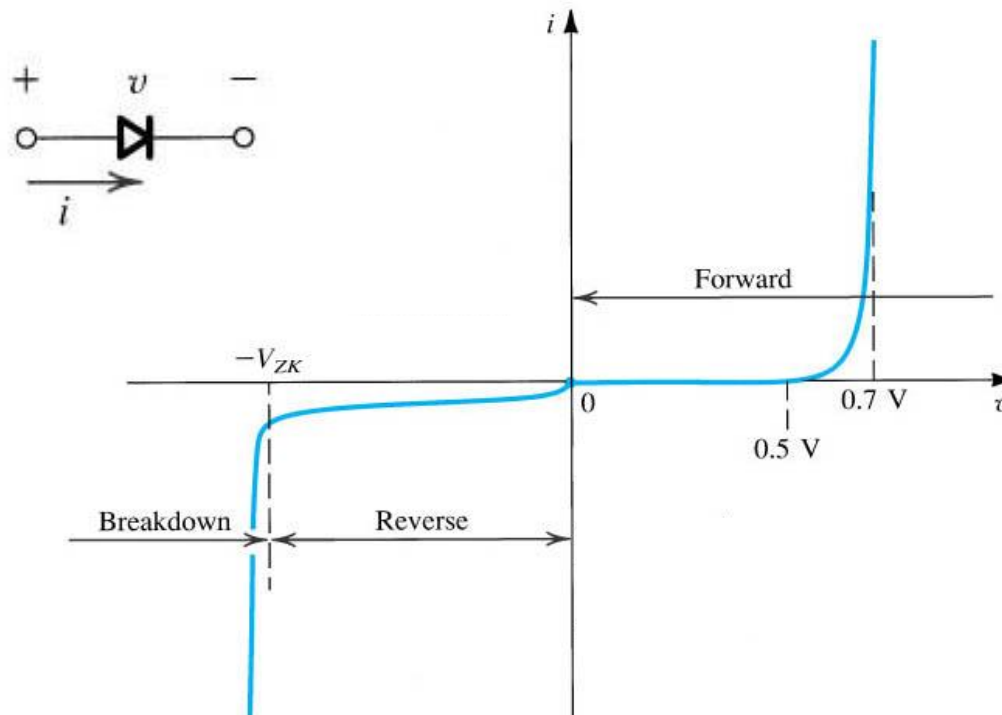
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Diodes

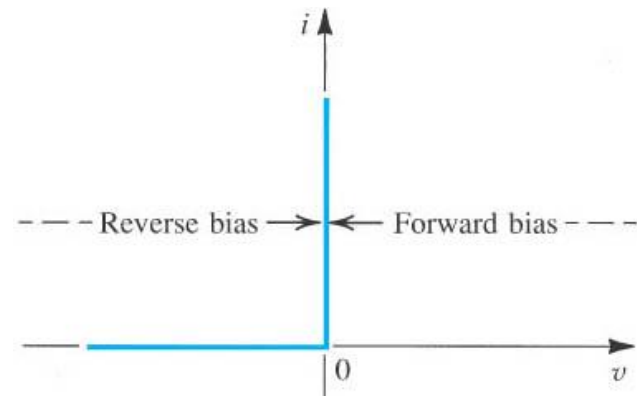
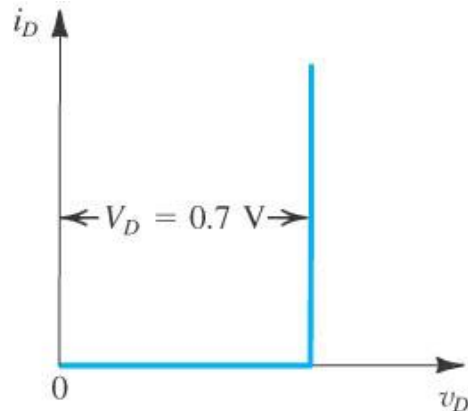
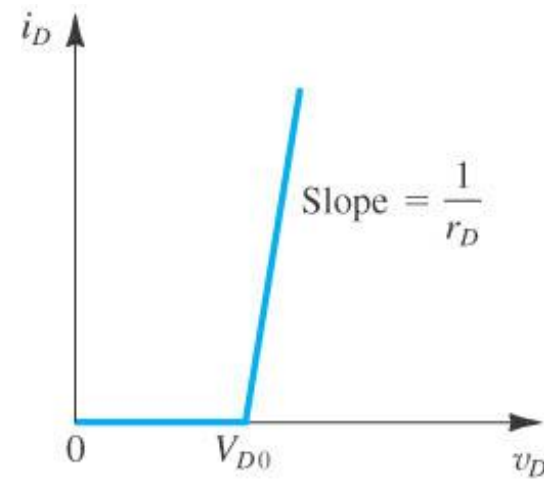
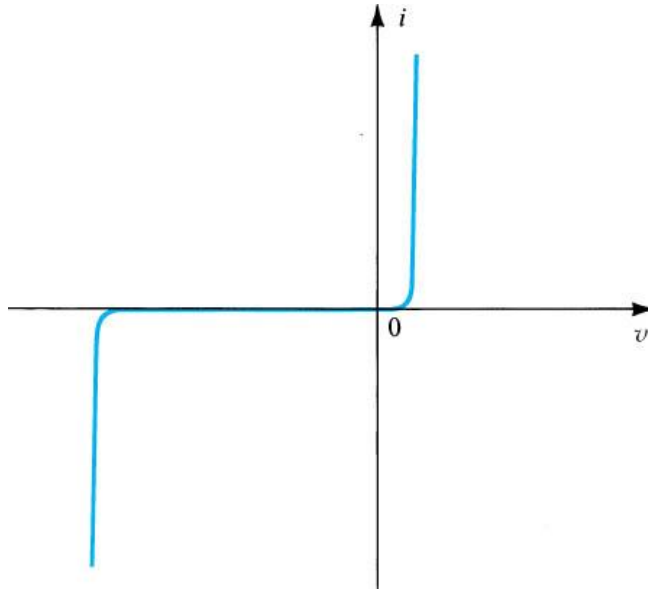
I-V characteristics

- A diode is merely a PN-junction (with terminals connected at both ends)
- Reminder from ELEC-H301
 - 3 regions: forward bias, reverse bias and breakdown



Diodes

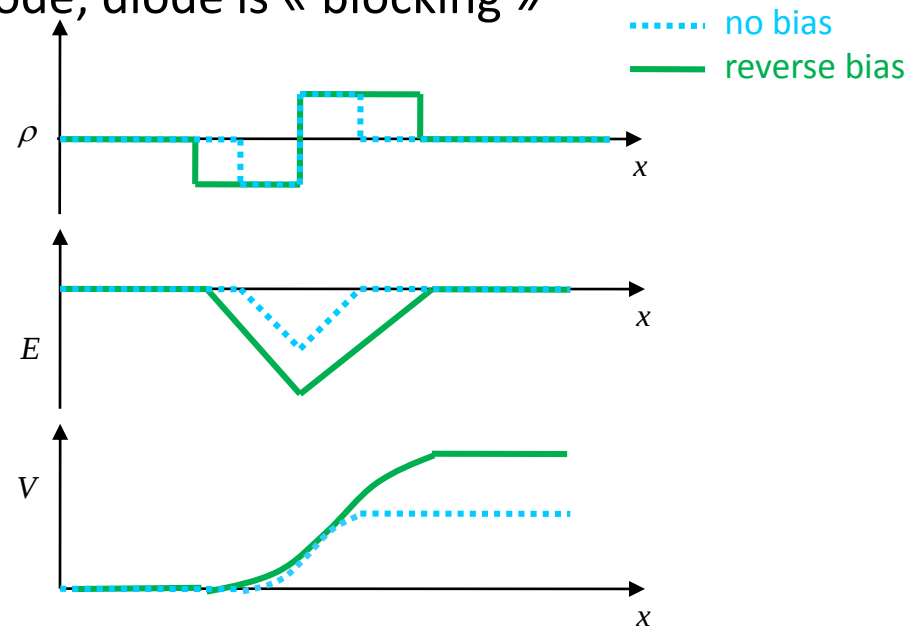
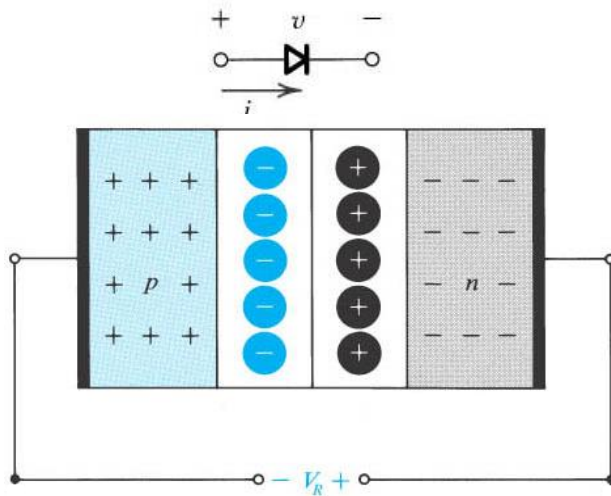
Several levels of simplification ...



Diodes: PN junction under reverse bias

The depletion region widens

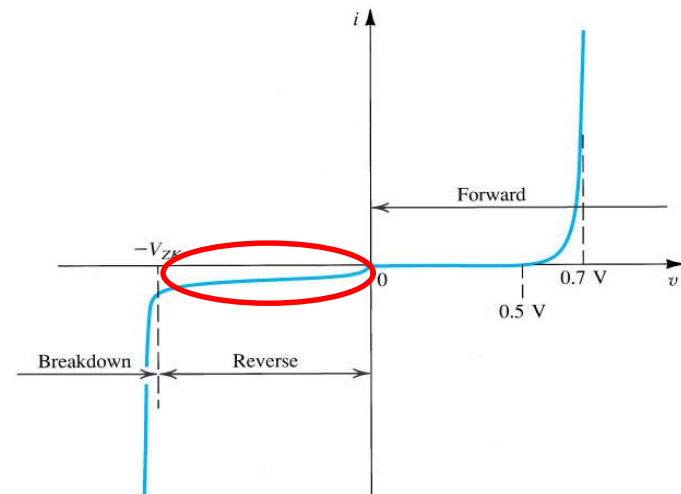
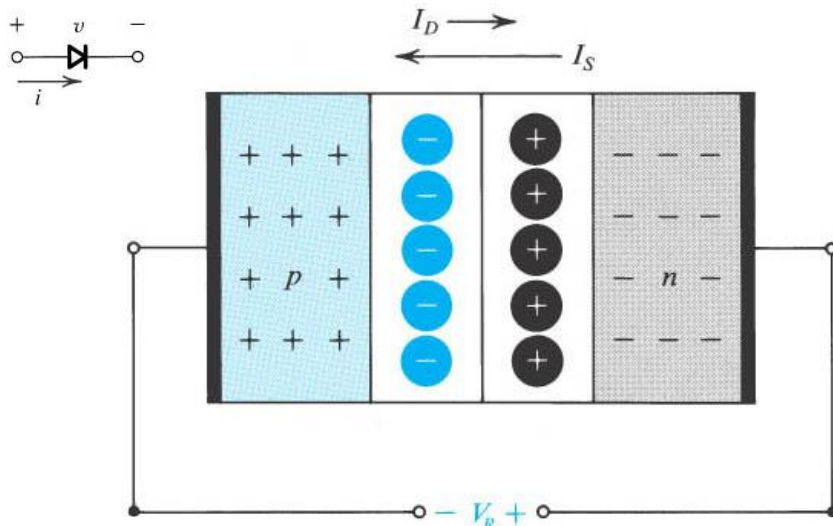
- Voltage source pulls electrons from n-region and holes from p-region around the depletion region
- Depletion region widens => barrier voltage V_0 increases
 - ⇒ It becomes even harder for holes/electrons from the p/n region to get through the depletion region =>
 - ⇒ no current goes through the diode; diode is « blocking »



Diodes: PN junction under reverse bias

Reverse current is due to minority carriers

- Since the barrier voltage of the depletion region increases, the diffusion current I_D (due to majority holes/electrons in the p/n region) will decrease
 - However, the diffusion current I_S (due to minority electrons/holes in the p/n region) will not change
- => a slight reverse current will appear, due to **minority carriers**



Diodes: PN junction in the breakdown region

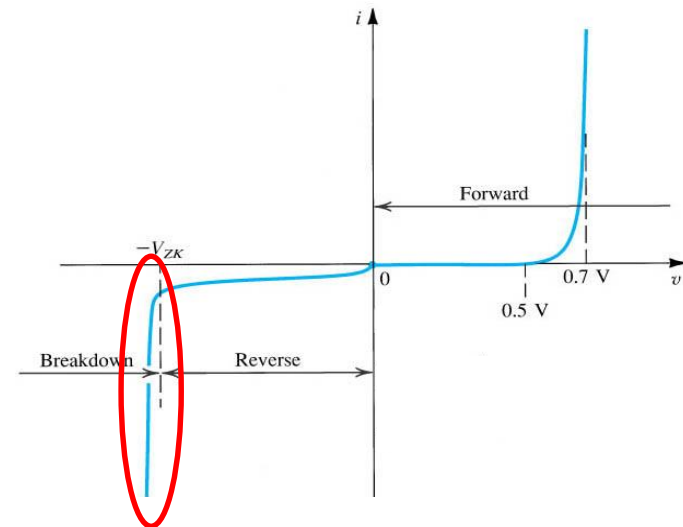
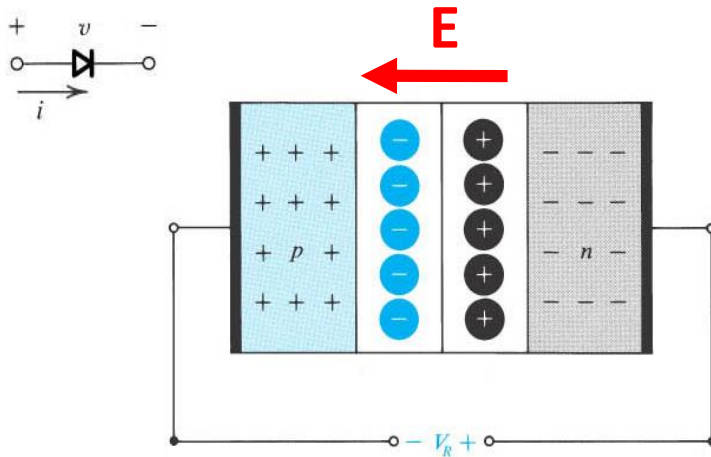
Reverse voltage keeps increasing

The electric field in the depletion region keeps increasing

1st effect:

⇒ E-field is able to break covalent bonds in the depletion region
(« Zenner effect ») , with negligible increase in junction voltage

⇒ Reverse current will increase

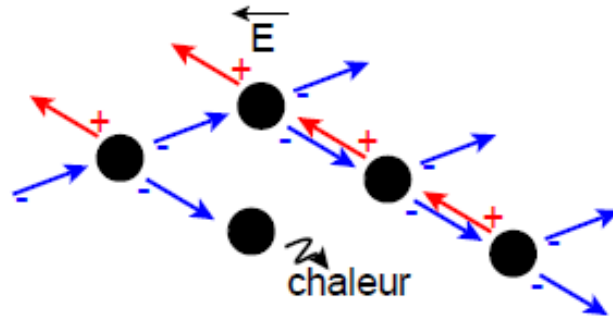


Diodes: PN junction in the breakdown region

Reverse voltage keeps increasing

The electric field in the depletion region keeps increasing

2nd effect:



- ⇒ Minority carriers going through depletion region have so much kinetic energy that they are able to break covalent bonds
- ⇒ Each liberated electron has sufficient energy to cause another ionizing collision (« Avalanche effect »)
- ⇒ reverse current increases fast

Diodes: PN junction in the breakdown region

PN breakdown is not a destructive process as such

- When decreasing the reverse voltage, diode will return to its original state

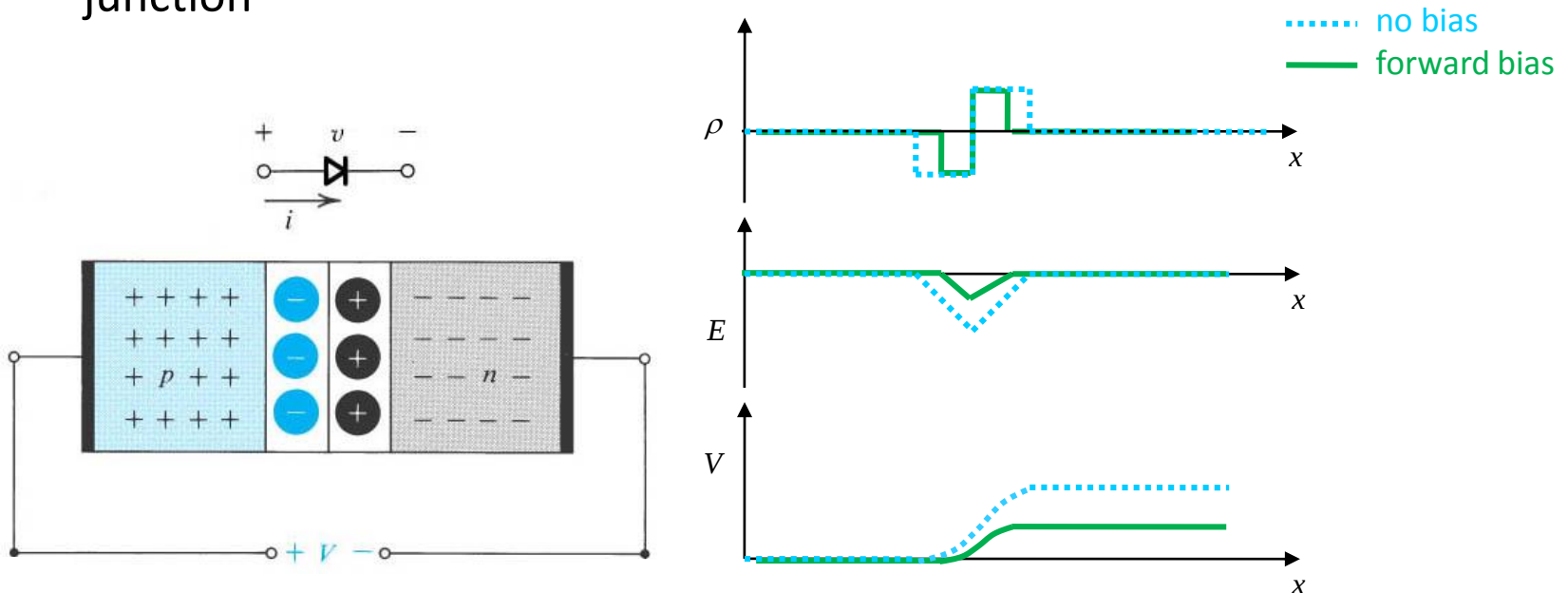
However... reverse current can cause high power dissipation

⇒ This power dissipation can degrade the PN junction irreversibly, maybe even cause the Silicon to fuse

Diodes: PN junction under forward bias

Depletion regions narrows

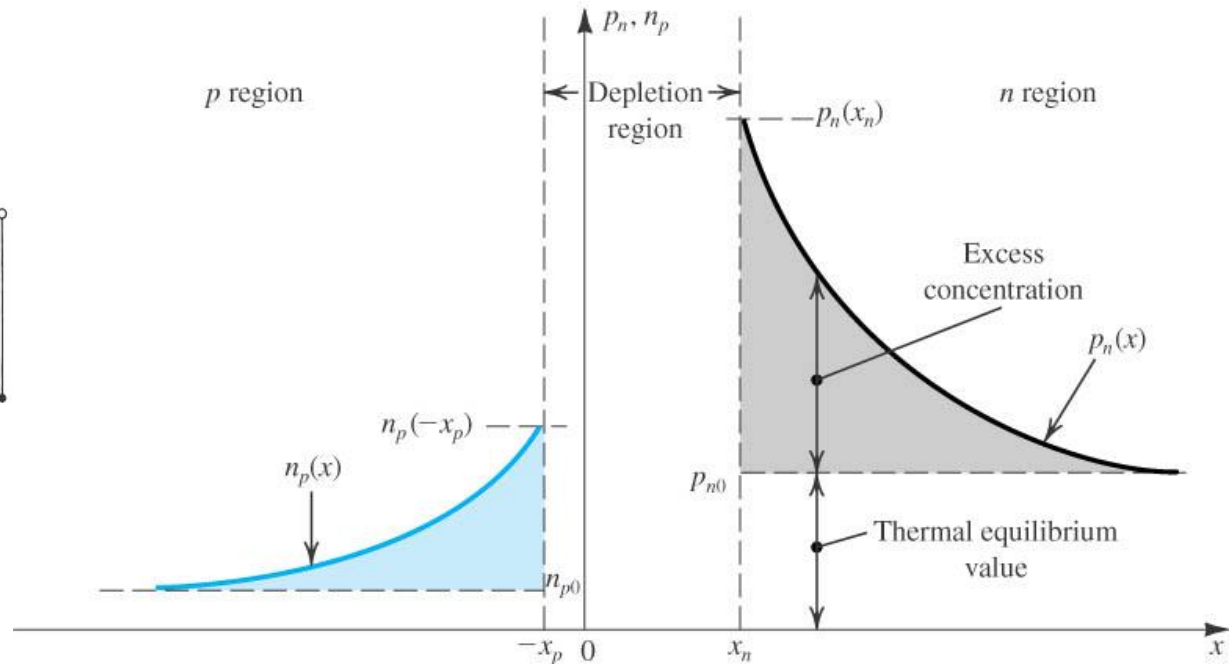
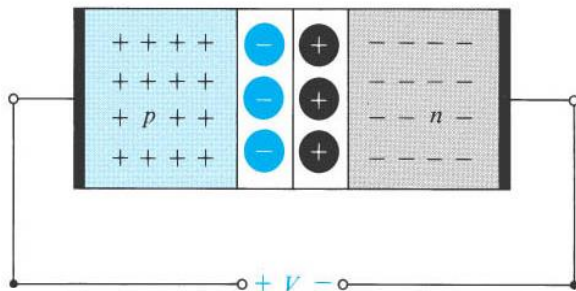
- Voltage source pushes holes in the p-region and electrons in the n-region closer to the depletion region
 - ⇒ Depletion region narrows
 - ⇒ Barrier voltage V_0 decreases
 - ⇒ Holes from p-region / electrons from n-region are injected accross the junction



Diodes: PN junction under forward bias

Minority carrier concentration increases

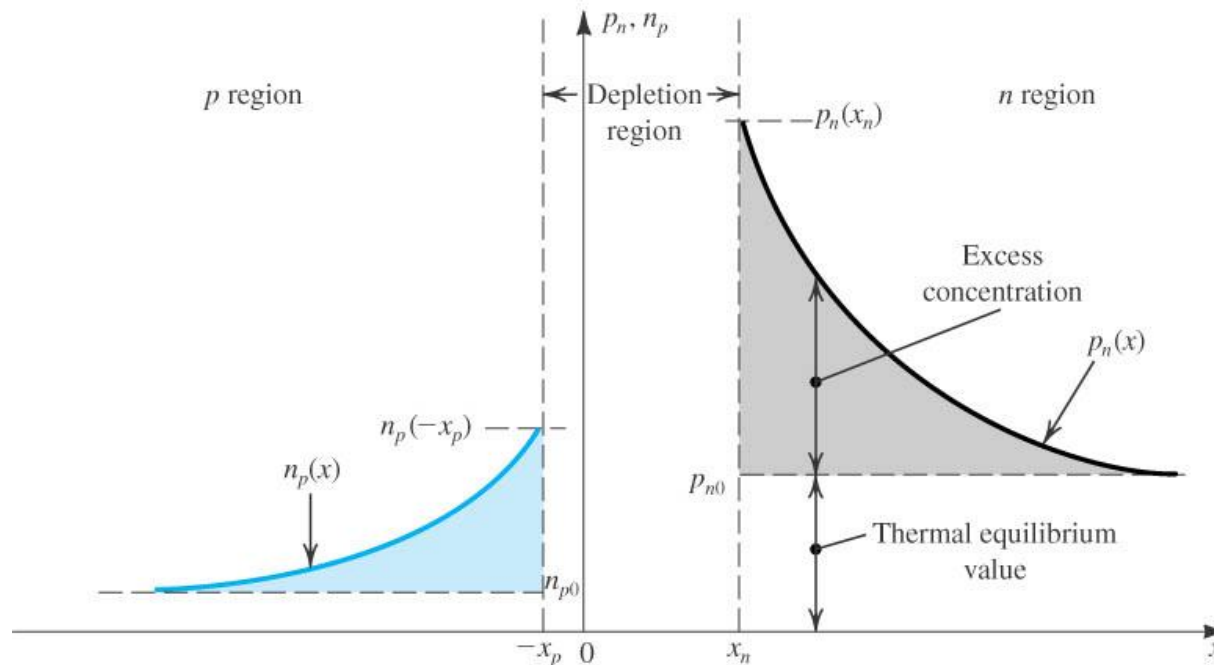
- Holes from p-region are injected accross the junction
 - ⇒ Increase minority carrier concentration in the n-region
 - ⇒ Holes recombine with majority electron in n-region
 - ⇒ Highest concentration of minority carriers close to junction
- Similar for electrons from the n-region



Diodes: PN junction under forward bias

Excess minority carrier concentration causes current

- Highest concentration of minority carriers close to junction
 $\Rightarrow$ from semiconductor physics: $p_n(x_n) = p_{n0}e^{V/V_T}$
 $\Rightarrow$ concentration of carriers: $p_n(x) = p_{n0} + [p_n(x_n) - p_{n0}]e^{-(x-x_n)/L_p}$



Diodes: PN junction under forward bias

Excess minority carrier concentration causes current

- Gradient in minority charge carriers causes diffusion current:

$$\begin{aligned} J_p &= qD_p \frac{dp_n(x)}{dx} \\ &= q \frac{D_p}{L_p} p_{n0} (e^{V/V_T} - 1) e^{-(x-x_n)/L_p} \end{aligned}$$

$$\text{at } x = x_n : J_p = q \frac{D_p}{L_p} p_{n0} (e^{V/V_T} - 1)$$

- Similarly, for minority electrons in p-region: $J_n = q \frac{D_n}{L_n} n_{p0} (e^{V/V_T} - 1)$

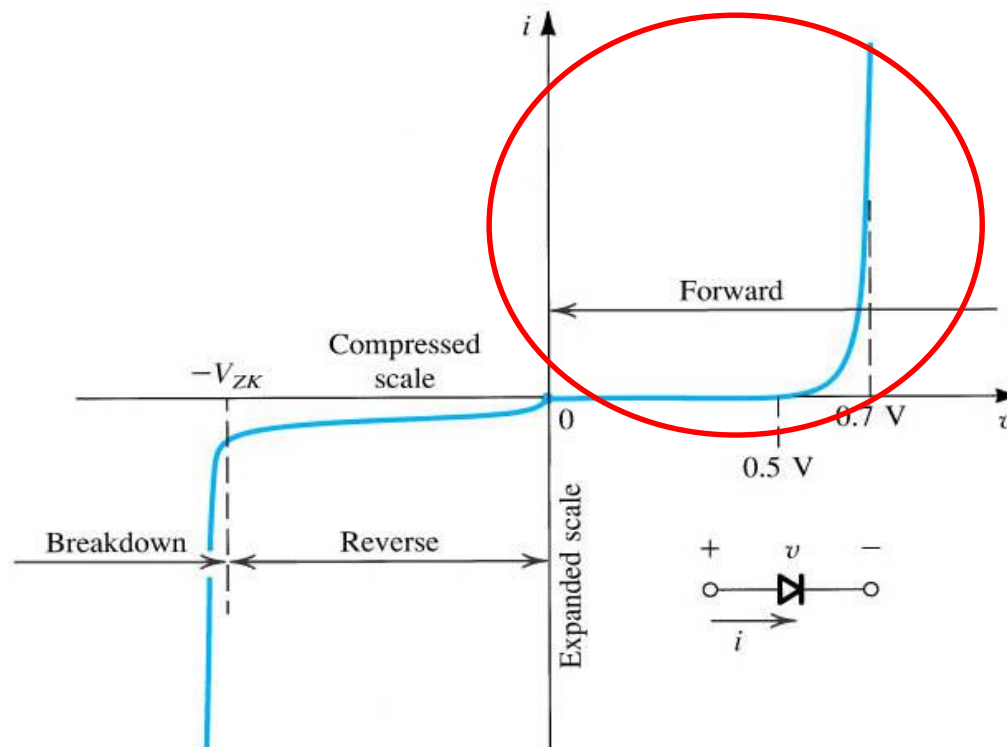
=> Total current for cross-section A :

$$I = A \underbrace{\left(q \frac{D_p}{L_p} p_{n0} + q \frac{D_n}{L_n} n_{p0} \right)}_{=I_S} (e^{V/V_T} - 1)$$

Diodes: PN junction under forward bias

Excess minority carrier concentration causes current

- Total current for cross-section A : $I = I_s (e^{V/V_T} - 1)$
 $\Rightarrow$ diode characteristic I-V equation in the forward bias region



Diodes: PN junction under forward bias

Excess minority carrier lifetime

- When turning of the forward voltage, it takes some time for the excess minority carriers to recombine with majority carriers: $p_n(x) = p_{n0} + \left([p_n(x_n) - p_{n0}] e^{-(x-x_n)/L_p} \right) e^{-t/\tau_p}$
- This excess minority carrier lifetime is linked to the diffusion length:

$$L_p = \sqrt{D_p \tau_p}$$

$$L_n = \sqrt{D_n \tau_n}$$

L_p ranges from 1μm to 100μm

τ_p ranges from 1 ns to 10000 ns

⇒ determines the speed of diode switching

⇒ τ_p can be reduced by increasing crystal purity, or by doping with Au or Pt

PN junctions: conclusions

Summary

- Forward bias
 - Depletion area narrows and more majority carriers cross junction
 - Cause excess minority carriers
 - ⇒ create diffusion current
 - ⇒ Diffusion current exponential with forward voltage
- Reverse bias
 - Depletion area widens and less majority carriers cross junction
 - Minority carriers still cross junction
 - ⇒ I_S (due to minority carriers) becomes larger than I_D (due to majority carriers)
 - ⇒ Small reverse current

PN junctions: conclusions

Summary

- Breakdown region
 - Zenner effect: large E-field breaks covalent bonds
⇒ Large reverse current
 - Large E-field: minority carriers have enough kinetic energy to break covalent bonds
⇒ Freed electrons also break other covalent bonds (Avalanche effect)
⇒ Large reverse current

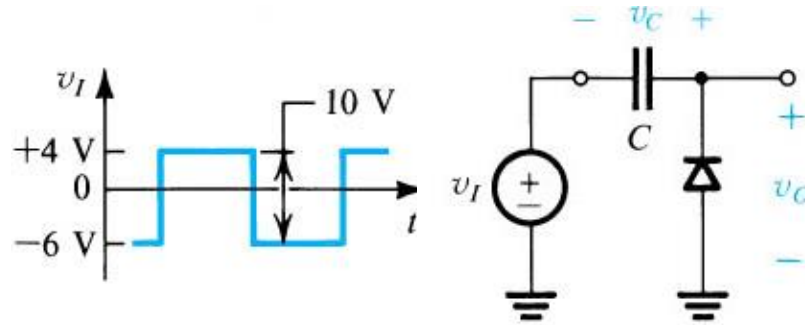
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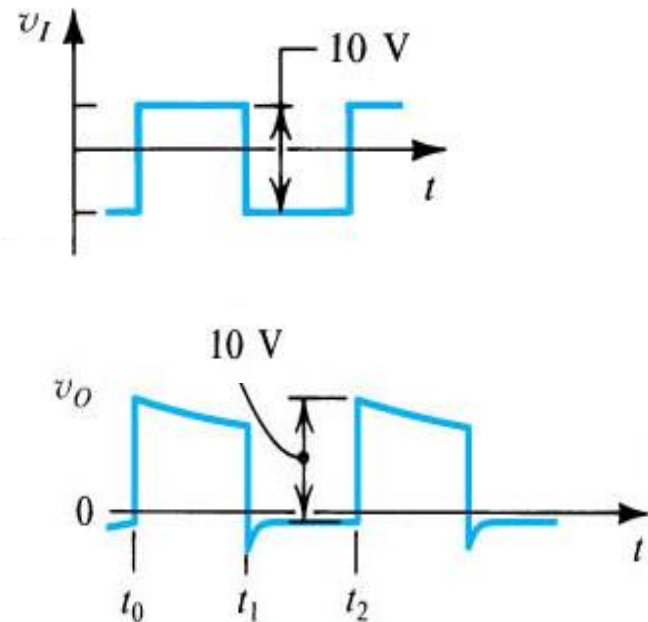
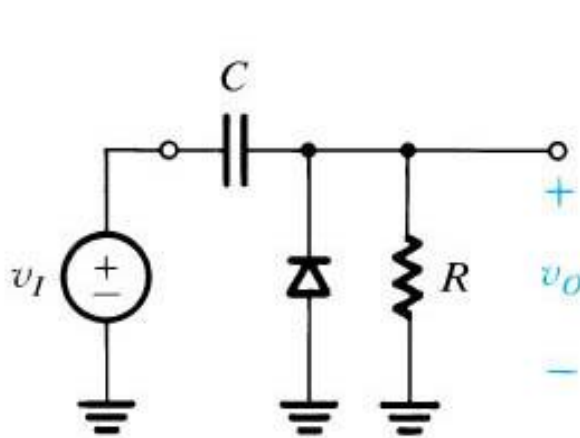
Circuits with diodes: DC restorer

adds a DC component to an AC signal



Circuits with diodes: DC restorer

When adding a load, capacitor discharges slightly

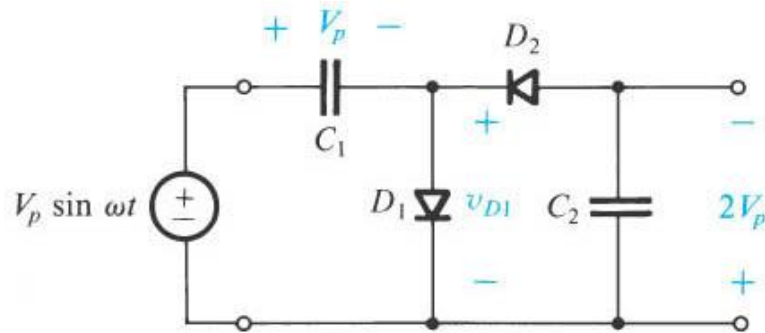


⇒ Choose RC constant wisely with respect to signal period!

Circuits with diodes: voltage doubler

Double AC voltage to DC voltage

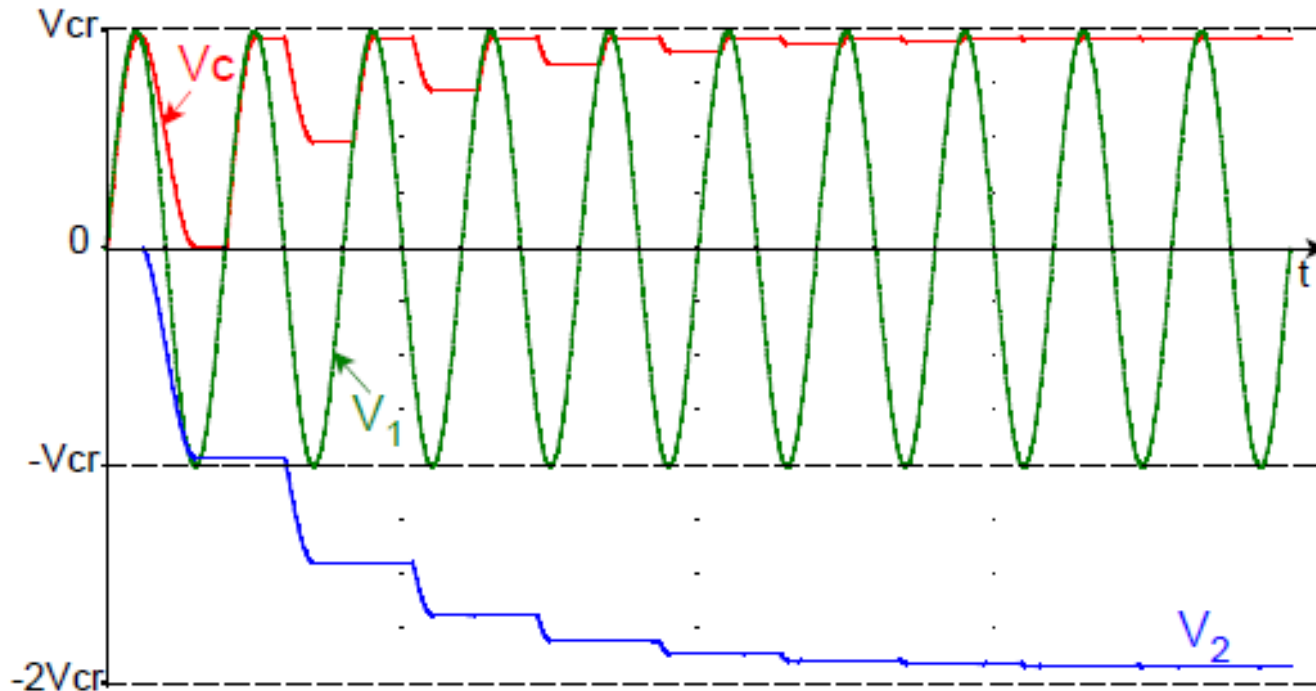
- Output is DC voltage twice the amplitude of AC input



Circuits with diodes: voltage doubler

PSpice simulation

- Capacitor takes a few cycles to be fully charged
- Threshold voltage diode causes slight voltage reduction



Circuits with diodes: voltage multiplier

previous circuit can be generalized

- V_{cr} is the amplitude of the source
- Every diode has to be able to sustain a reverse voltage of $2V_{cr}$ without going into avalanche

